D OR PW PACKAGE

SLVS004D - APRIL 1979 - REVISED JANUARY 2002

- **Output Voltage Range Adjustable From** 1.2 V to 32 V When Used With an External **Resistor Divider**
- **Output Current Capability of 100 mA**
- Input Regulation Typically 0.01% Per **Input-Voltage Change**
- **Output Regulation Typically 0.5%**
- Ripple Rejection Typically 80 dB

three-terminal regulators.

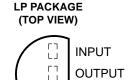
# description

The TL317 is an adjustable three-terminal positive-voltage regulator capable of supplying 100 mA over an output-voltage range of 1.2 V to 32 V. It is exceptionally easy to use and requires only two external resistors to set the output voltage.

In addition to higher performance than fixed

(TOP VIEW) **INPUT** Π NC OUTPUT [ OUTPUT 6 OUTPUT OUTPUT [ **ADJUSTMENT** □ NC 5

NC - No internal connection OUTPUT terminals are all internally connected.



**ADJUSTMENT** 

regulators, this regulator offers full overload protection available only in integrated circuits. Included on the chip are current-limiting and thermal-overload protection. All overload-protection circuitry remains fully functional, even when ADJUSTMENT is disconnected. Normally, no capacitors are needed unless the device is situated far from the input filter capacitors, in which case an input bypass is needed. An optional output capacitor can be added to improve transient response. ADJUSTMENT can be bypassed to achieve very high ripple rejection, which is difficult to achieve with standard

In addition to replacing fixed regulators, the TL317 regulator is useful in a wide variety of other applications. Since the regulator is floating and sees only the input-to-output differential voltage, supplies of several hundred volts can be regulated as long as the maximum input-to-output differential is not exceeded. Its primary application is that of a programmable output regulator, but by connecting a fixed resistor between ADJUSTMENT and OUTPUT, this device can be used as a precision current regulator. Supplies with electronic shutdown can be achieved by clamping ADJUSTMENT to ground, programming the output to 1.2 V, where most loads draw little current.

The TL317C is characterized for operation over the virtual junction temperature range of 0°C to 125°C.

### **AVAILABLE OPTIONS**

	PACKAGED DEVICES			
ТЈ	SMALL OUTLINE (D)	PLASTIC (LP)	PLASTIC THIN SHRINK SMALL OUTLINE (PW)	
0°C to 125°C	TL317CD	TL317CLP	TL317CPW	

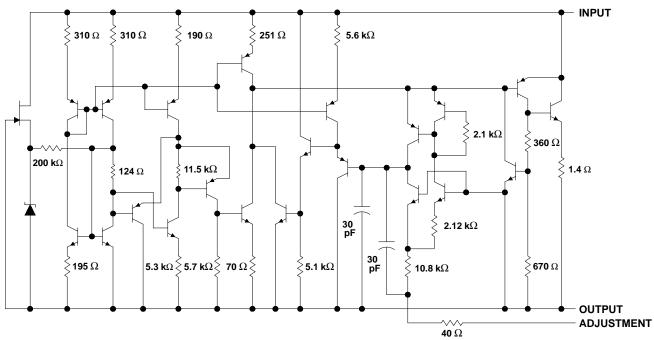
The D and LP packages are available taped and reeled. Add the suffix R to device type (e.g., TL317CDR). The PW package is only available taped and reeled.



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### schematic



NOTE A: All component values shown are nominal.

# absolute maximum ratings over operating temperature range (unless otherwise noted)†

Input-to-output differential voltage, V <sub>I</sub> – V <sub>O</sub>	
Package thermal impedance, $\theta_{JA}$ (see Notes 1 and 2): D pack	age 97°C/W
LP pag	kage156°C/W
PW pa	ckage 149°C/W
Lead temperature 1,6 mm (1/16 inch) from case for 10 second	s 260°C
Storage temperature range, T <sub>stg</sub>	–65°C to 150°C

<sup>†</sup> Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

## recommended operating conditions

			MIN	MAX	UNIT
V <sub>I</sub> – V <sub>O</sub> Input-to-output voltage differential			35	V	
lo	Output current		2.5	100	mA
TJ	Operating virtual-junction temperature	TL317C	0	125	°C



NOTES: 1. Maximum power dissipation is a function of  $T_J(max)$ ,  $\theta_{JA}$ , and  $T_A$ . The maximum allowable power dissipation at any allowable ambient temperature is  $P_D = (T_J(max) - T_A)/\theta_{JA}$ . Operating at the absolute maximum  $T_J$  of 150°C can affect reliability.

<sup>2.</sup> The package thermal impedance is calculated in accordance with JESD 51-7.

SLVS004D - APRIL 1979 - REVISED JANUARY 2002

# electrical characteristics over recommended operating virtual-junction temperature range (unless otherwise noted)

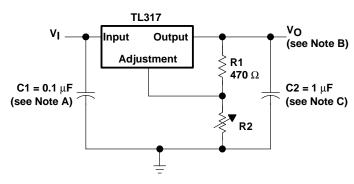
DADAMETED	TEST CONDITIONS <sup>†</sup>		TL317C			LINUT
PARAMETER			MIN	TYP	MAX	UNIT
Input voltage regulation (see Note 3)	V: V = = 5 V +0 25 V	$T_J = 25^{\circ}C$		0.01	0.02	%V
input voltage regulation (see Note 3)	$V_{I} - V_{O} = 5 \text{ V to } 35 \text{ V}$	$I_O = 2.5 \text{ mA to } 100 \text{ mA}$		0.02	0.05	
	$V_0 = 10 \text{ V}, \qquad f = 120 \text{ Hz}$			65		dB
Ripple regulation	V <sub>O</sub> = 10 V, 10-μF capacitor between ADJUSTMENT and ground		66	80		
	$V_I = 5 \text{ V to } 35 \text{ V},$ $I_O = 2.5 \text{ mA to } 100 \text{ mA},$ $T_J = 25^{\circ}\text{C}$	$V_O \le 5 V$		25		mV
Output voltage regulation		$V_O \ge 5 \text{ V}$		5		mV/V
	$V_I = 5 \text{ V to } 35 \text{ V},$ $I_O = 2.5 \text{ mA to } 100 \text{ mA}$	$V_O \le 5 V$		50		mV
		$V_O \ge 5 V$		10		mV/V
Output voltage change with temperature	T <sub>J</sub> = 0°C to 125°C			10		mV/V
Output voltage long-term drift After 1000 hours at $T_J = 125^{\circ}\text{C}$ and $V_I - V_O = 35 \text{ V}$			3	10	mV/V	
Output noise voltage	f = 10 Hz to 10 kHz,	T <sub>J</sub> = 25°C		30		μV/V
Minimum output current to maintain regulation	$V_I - V_O = 35 V$			1.5	2.5	mA
Peak output current	$V_I - V_O \le 35 \text{ V}$		100	200		mA
ADJUSTMENT current				50	100	μА
Change in ADJUSTMENT current	$V_I - V_O = 2.5 \text{ V to } 35 \text{ V},$	$I_{O} = 2.5 \text{ mA to } 100 \text{ mA}$		0.2	5	μΑ
Reference voltage (output to ADJUSTMENT)	$V_I - V_O = 5 \text{ V to } 35 \text{ V},$ P \le rated dissipation	$I_O = 2.5 \text{ mA to } 100 \text{ mA},$	1.2	1.25	1.3	V

<sup>†</sup> Unless otherwise noted, these specifications apply for the following test conditions:  $V_I - V_O = 5$  V and  $I_O = 40$  mA. Pulse-testing techniques must be used that maintain the junction temperature as close to the ambient temperature as possible. All characteristics are measured with a 0.1- $\mu$ F capacitor across the input and a 1- $\mu$ F capacitor across the output.

NOTE 3: Input voltage regulation is expressed here as the percentage change in output voltage per 1-V change at the input.



### APPLICATION INFORMATION



NOTES: A. Use of an input bypass capacitor is recommended if regulator is far from the filter capacitors.

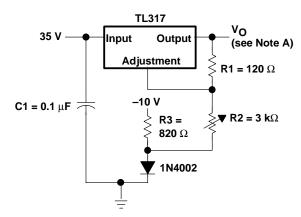
B. Output voltage is calculated from the equation:

$$V_O = V_{ref} \left( 1 + \frac{R2}{R1} \right)$$

where:  $V_{\text{ref}}$  equals the difference between OUTPUT and ADJUSTMENT voltages ( $\approx$ 1.25 V).

C. Use of an output capacitor improves transient response but is optional.

Figure 1. Adjustable Voltage Regulator

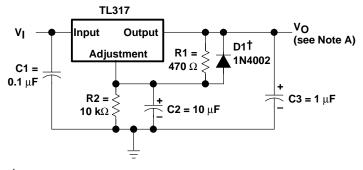


NOTE A: Output voltage is calculated from the equation:

$$V_{O} = V_{ref} \left( 1 + \frac{R2 + R3}{R1} \right) - 10 \text{ V}$$

where:  $V_{\text{ref}}$  equals the difference between OUTPUT and ADJUSTMENT voltages ( $\approx$ 1.25 V).

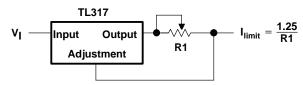
Figure 2. 0-V to 30-V Regulator Circuit



†D1 discharges C2 if output is shorted to ground.

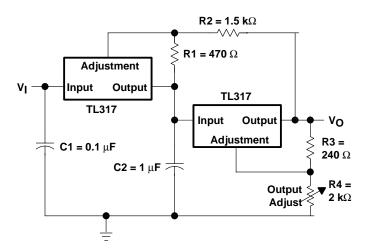
NOTE A: Use of an output capacitor improves transient response but is optional.

Figure 3. Regulator Circuit With Improved Ripple Rejection



**Figure 4. Precision Current-Limiter Circuit** 

## **APPLICATION INFORMATION**



**Figure 5. Tracking Preregulator Circuit** 

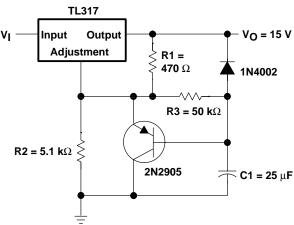


Figure 6. Slow Turnon 15-V Regulator Circuit

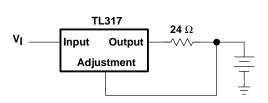


Figure 7. 50-mA Constant-Current Battery-Charger Circuit

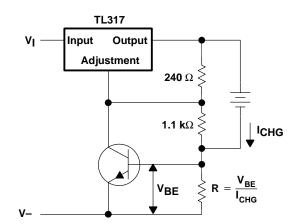
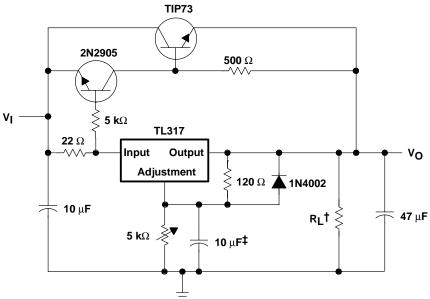


Figure 8. Current-Limited 6-V Charger

# **APPLICATION INFORMATION**



<sup>†</sup> Minimum load current is 30 mA.

Figure 9. High-Current Adjustable Regulator



 $<sup>\</sup>mbox{\ensuremath{\mbox{\footnotesize 1}}}$  Optional capacitor improves ripple rejection.

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